

Communication Quality Estimation Observer: An Approach for Integrated Communication Quality Estimation and Control for Digital-Twin-Assisted Cyber-Physical Systems

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SUMMARY Cyber-physical systems (CPSs) assisted by digital twins (DTs) integrate sensing-actuation loops over communication networks in various infrastructure services and applications. This study overviews the concept, methodology, and applications of the integrated communication quality estimation and control for the DT-assisted CPSs from both communications and control perspectives. The DT-assisted CPSs can be considered as networked control systems (NCSs) with virtual dynamic models of physical entities. A communication quality estimation observer (CQEO), which is an extended version of the communication disturbance observer (CDOB) utilized for time-delay compensation in NCSs, is proposed to estimate the integrated effects of the quality of services (QoS) and cyberattacks on the NCS applications. A path diversity technique with the CQEO is also proposed to achieve reliable NCSs. The proposed technique is applied to two kinds of NCSs: remote motor control and haptic communication systems. Moreover, results of the simulation on a haptic communication system show the effectiveness of the proposed approach. In the end, future research directions of the CQEO-based scheme are presented.

key words: *communication quality, networked control system, disturbance observer, digital twin, cyber-physical system*

1. Introduction

A cyber-physical system (CPS) only functions if information, communications, and control technologies are integrated [1], [2]. In particular, sensing and actuation over communication networks are key techniques for building more sophisticated and diversified CPSs. From a control perspective, the CPS can be considered as a networked control system (NCS) [3]–[5], in which controllers, sensors, and actuators are connected through communication networks. Therefore, the system performance and stability are significantly affected by network-induced constraints such as time delays, packet losses, sampling intervals, data quantization, and network security [6], [7]. These network-induced imperfections in the NCSs have stimulated multidisciplinary studies on information, communications, and control engineering in recent years [8]–[12].

To reduce the time delays for delay-sensitive CPS applications, such as teleoperation, augmented reality, serious gaming, and smart grid, the concept of the tactile internet (TI) has been proposed from a communications perspective

[13]. TI is aimed at achieving a 1-ms end-to-end latency, for example, by using multi-access edge computing (MEC) [14]. As for wired networks, various edge computing architectures based on passive optical networks (PONs) have been proposed [15]–[17] for medical and robotic applications. Maier et al. [18] proposed a TI architecture using fiber wireless (FiWi)-enhanced mobile networks for a teleoperation system. The TI is expected to become an ultra reliable and low latency communications (URLLC) service in the beyond-5G or 6G era [19].

The TI will enable the ultra-low-latency CPS services by using the MEC-based architecture. However, the MEC-based architecture assumes the locality of distributed physical entities. In fact, it is difficult to achieve cooperative control of devices distributed over a wide area and teleoperation from distant locations. Therefore, in addition to adopting a TI-based approach, application and network control techniques to overcome the network constraints such as time delays in CPSs are necessary. The network-level communication quality can be estimated as the quality of services (QoS), for example, latency, packet loss rate, and throughput. Park et al. [20] proposed a robust multipath selection algorithm based on the round-trip time (RTT) for a CPS. Furthermore, various QoS-based network control and data transmission schemes for NCSs have been discussed [21]–[23]. At the application level, the QoS information has been also utilized to compensate for the effects of time delays in NCSs [24]–[26].

Although the QoS-based control schemes can theoretically achieve robust and reliable NCSs, the measurement and estimation of each QoS metric are often difficult. Therefore, it is worth estimating and controlling the integrated communication quality at the application level in NCSs. A digital twin (DT) has recently emerged as a promising tool to estimate the integrated communication quality. The DT is a virtual model of a physical system and can be used to predict, optimize, and control future behavior of entities in the system [27]–[29]. Newrzella et al. [30] classified the DT concept across industries, in which there have been various definitions of the DT. The DT-assisted CPSs have been proposed for various applications, including the manufacturing system [31], on-orbit spacecraft [32], and smart grids [33]. The DT concept covers not only model-based approach but also data-driven approach in that modeling errors can be observed by real-time communication between a virtual model

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and a physical entity. Moreover, not only an analytical model of a physical entity but also a simulator can be utilized as a DT in complicated systems that are difficult to model, such as multi-agent systems. In that sense, this paper uses the term of DT-assisted CPS while a simplified analytical model is utilized for illustration purpose in this study. In the DT-assisted CPSs, the effects of the network-induced constraints can be mitigated by application and network control based on the integrated communication quality at the application level if the system modeling is possible, for example, by using sophisticated machine learning techniques.

Natori et al. [34], [35] proposed a communication disturbance observer (CDOB) to compensate for the effects of time delays on NCSs. The CDOB does not need any time-delay models in the system and is effective for not only constant but also time-varying delays. The CDOB estimates the effects of time delays as a disturbance at the application level by using a nominal model of a controlled subsystem, which can be considered as a simplified DT. A double disturbance observer (DDOB) was proposed to cope with the modeling errors in the CDOB-based time-delay compensation [36], [37]. The CDOB-based NCSs are affected by not only time delays but also packet losses [38]. The effects of packet losses on NCSs can be also considered as a disturbance [39]. The CDOB can estimate the integrated communication quality, including both time delays and packet losses as a disturbance [40], [41]. As for the NCSs with the CDOB, the steady-state errors caused by time-varying delays and packet losses have been discussed [42], [43]. In addition, a modified CDOB was proposed to compensate for the effects of data losses caused by sleep-based energy-efficient network interfaces [44]. Moreover, Lee et al. [45] proposed a fuzzy-based CDOB architecture based on neural network modeling. In this architecture, a learning algorithm is utilized to perform system identification.

The above-mentioned CDOBs involve techniques for application control in NCSs. The CDOB is also effective for network control, for example, for path selection in multipath routing. The disturbance estimated by the CDOB includes the effects of time delays, jitter, packet loss, etc. This means that the disturbance indicates the integrated communication quality at the application level [46]. However, the integrated communication quality estimation based on the CDOB has not been applied to network control. Recently, cybersecurity in NCSs has been recognized as a critical issue [47]. Various attack scenarios, such as false data injection (FDI) attacks, replay attacks, and denial-of-service (DoS) attacks, have been studied [48]. The effects of FDI-based cyberattacks against the NCSs can be also considered as disturbance. To detect and mitigate data tampering attacks in the NCSs, a tamper detection observer (TDO) was proposed [49], [50]. The TDO includes an algorithm to select a path that has not been attacked out of multiple paths based on observer-based disturbance estimation. The communication quality estimation through the CDOB as well as TDO should be applied to network control, and in this study, we discuss the integrated communication quality, including the effects of both QoS

and cybersecurity.

The observer-based approach to estimating the integrated communication quality has been applied to various NCSs. For example, a haptic communication system is one of the interactive NCSs [51]. Master and slave haptic devices are bilaterally controlled over networks. To compensate for the effects of time delays, the configuration of the CDOB for the bilateral control was discussed in [52]. In addition, the configuration of the TDO was discussed in [53] and [54] to detect and mitigate data tampering attacks. Moreover, network control techniques based on the QoS for haptic communication systems have been studied. Furthermore, multipath routing techniques were proposed to improve the reliability of the system in [55] and [56]. A server selection technique based on the QoS information was also proposed in [57]. In haptic communication systems, however, the integrated communication quality estimation based on the CDOB has not been applied to network control.

This study overviews the concept, methodology, and applications of the integrated communication quality estimation and control for the DT-assisted CPSs from both communications and control perspectives. A communication quality estimation observer (CQEO), which is an extended version of the CDOB, is proposed to estimate the integrated effects of the QoS and cyberattacks on various NCS applications. The CQEO is configured with multiple network paths and a nominal model of a controlled subsystem. We also propose a CQEO-based network control technique using path diversity to achieve reliable NCSs. Figure 1 summarizes the proposed approach, more traditional approaches, and the CDOB-based approach. The proposed CQEO is applied to two types of NCSs: remote motor control and haptic communication systems. We perform simulations on the haptic communication system; the results confirm the effectiveness of the proposed technique.

This rest of the paper is organized as follows: The immediately following section describes the concept of the DT-assisted CPSs and its modeling. Section 3 describes the communication quality estimation and control schemes using the CDOB and CQEO. Section 4 describes the imple-

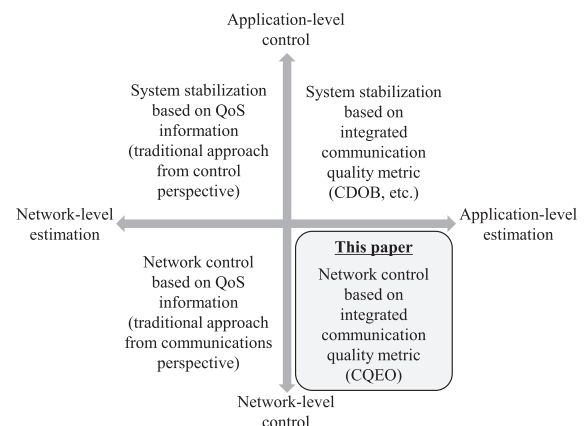


Fig. 1 Classification of communication quality estimation and control.

mentation of the CQEO into the path diversity technique for client-server NCSs, taking a remote motor control system as an example. Section 5 describes the implementation of the CQEO into the path diversity technique for peer-to-peer NCSs, taking a haptic communication system as an example. The results of the simulation on the haptic communication system are shown in Sect. 6. Finally, in Sect. 7, we present our conclusions and directions for future research.

2. DT-Assisted CPS and Its Modeling

This section describes the concept of the DT-assisted CPS, its modeling as a CPS service, and a disturbance observer (DOB) for disturbance rejection.

2.1 Concept of DT-Assisted CPS

The concept of the DT-assisted CPSs are illustrated in Fig. 2. The DT-assisted CPSs can be broadly classified into client-server and peer-to-peer CPSs. In the client-server CPS, as shown in Fig. 2(a), the controllers are located on the network-side cyber space, and the physical subsystems including sen-

sors and actuators are controlled over the networks. The DTs of the controlled subsystems are generated in the centralized or distributed servers. In addition, the local DTs are generated in the distributed devices including sensors and actuators. In the peer-to-peer CPS, as shown in Fig. 2(b), the controllers are located on the local cyber space, and the distributed devices including sensors and actuators are controlled multilaterally over the networks. Moreover, the DTs of peer subsystems are generated in the distributed devices.

The DT-assisted CPSs enable the following three functions to control the communication quality without direct QoS measurement at the network level. First, model-based predictive QoS compensation techniques can be implemented at the application level to stabilize the system. For example, the CDOB estimates and compensates for the effects of time delays by using the DT-based prediction. Second, multipath routing techniques can be implemented at the network level to achieve an appropriate path selection. For example, the CQEO-based network control technique using path diversity estimates the integrated communication quality in each path by using the DT-based prediction, and selects the path with the best communication quality. Third, controller allocation techniques can be implemented at the network level to achieve an appropriate server selection. For example, the CQEO-based network control technique estimates the integrated communication quality for each server by using the DT-based prediction, and selects the server with the best communication quality.

This study mainly focuses on the CQEO-based network control technique using path diversity for the client-server and peer-to-peer NCSs. A remote motor control system and a haptic communication system are taken as examples of the client-server and peer-to-peer NCSs, respectively.

2.2 DT-Assisted Networked Control as a CPS Service

If the DT-assisted feedback control over communication networks is considered as one of the CPS services, the system can be illustrated as Fig. 3. The feedback controller and physical plants, that is, the controlled subsystem including a sensor and an actuator, are connected over the forward and feedback networks. A proportional-integral-derivative (PID)-based controller can be implemented as the feedback

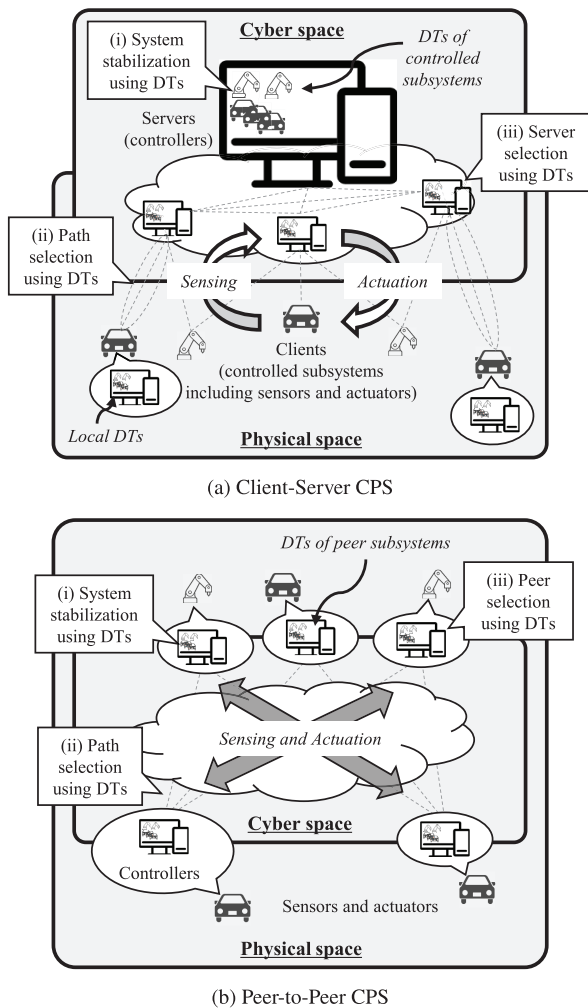


Fig. 2 Concept of the DT-assisted CPSs.

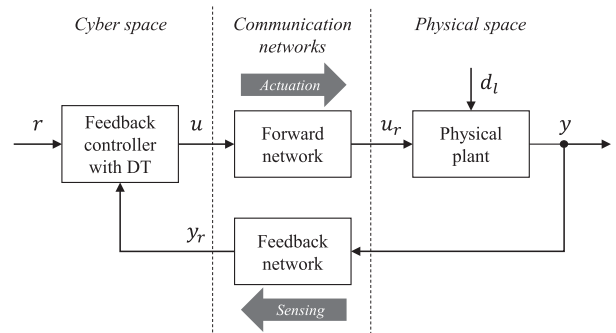


Fig. 3 Configuration of the DT-assisted NCS.

controller. The controller is designed based on the DT, that is, a model of the physical plant, which can be updated static or dynamically. In Fig. 3, r , u , and y denote the reference, control input, and response, respectively. The subscript r denotes the signal received over a network, where the packetized signal can be delayed or lost.

In general, the local disturbance d_l including modeling errors is input to the control system. In a motor control system, for example, the disturbance includes the inertia fluctuation, frictional torque, and load torque, which cannot be completely modelled in advance. As local disturbance affects not only the system stability and performance but also communication quality estimation as will be discussed hereafter, it becomes important to completely reject the local disturbance so that the physical plant works the same way as the DT.

2.3 DT-Assisted Disturbance Rejection Using DOB

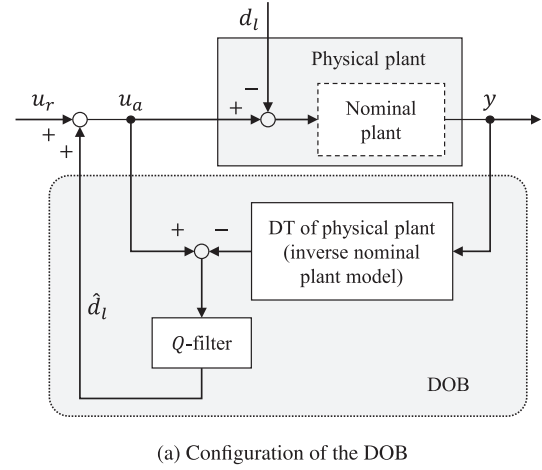
The DT of the physical plant has modeling errors even if the DT is updated based on real-time machine learning. The DOB estimates the system uncertainties or modeling errors, as disturbance d_l . The DOB has been utilized to improve the robustness against the disturbance in various practical control systems [58], [59]. The nominal model of the physical plant in the DOB can be considered as a DT. The merit of using the DOB lies in the simple implementation compared with other robust control schemes.

The overview of the DT-assisted disturbance rejection using the DOB is shown in Fig. 4. As shown in Fig. 4(a), the DOB comprises the DT and Q -filter on the plant side. The DT on the plant side, that is, local DT, is the inverse nominal model of the physical plant, which means the physical plant model without the disturbance. The Q -filter is a low-pass filter (LPF) to stabilize the system. The DOB calculates the estimated local disturbance \hat{d}_l through the Q -filter and compensates for the disturbance. The compensated signal u_a is input to the physical plant. The block diagram of Fig. 4(a) can be equivalently transformed into Fig. 4(b). Therefore, disturbance d_l is input to the system through the $(1 - Q)$ -filter, that is, high-pass filter (HPF), by the compensation using the DOB. The Q -filter with a higher cut-off frequency achieves more robust control against the disturbance. In practice, however, increasing the cut-off frequency is limited by system noises.

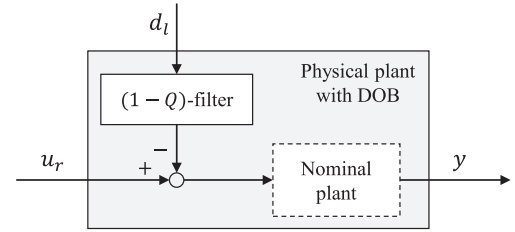
For example, we consider the transfer function of a linear time-invariant (LTI) system, $G_n(s)$, where s denotes the Laplace variable, as a nominal plant. The estimated local disturbance \hat{d}_l can be calculated as (1)

$$\begin{aligned} \hat{d}_l &= Q(s) \left(u_a - G_n(s)^{-1} y \right) \\ &= \frac{g_{dob}}{s + g_{dob}} d_l, \end{aligned} \quad (1)$$

where $Q(s)$ and g_{dob} denote the first-order Q -filter and its cut-off frequency, respectively. The output of the system is represented as (2)



(a) Configuration of the DOB



(b) Equivalent block diagram of Fig. 4(a)

Fig. 4 DT-assisted disturbance rejection using the DOB.

$$\begin{aligned} y &= G_n(s)u_r - (1 - Q(s))d_l \\ &= G_n(s)u_r - \frac{s}{s + g_{dob}}d_l. \end{aligned} \quad (2)$$

If the cut-off frequency g_{dob} is sufficiently large, the local disturbance d_l is completely rejected. Therefore, the physical plant works the same way as the local DT or nominal plant by using the DOB. Introducing the DOB into the local subsystem enables the integrated communication quality estimation and control as will be described in the following sections.

3. Communication Quality Estimation and Control

This section describes the CDOB-based application control technique and proposed CQEO-based network control technique for the DT-assisted CPSs. These techniques are discussed based on the DT-assisted NCS shown in Fig. 3.

3.1 CDOB-Based Estimation and Application Control

The CDOB is a passive method to measure the effects of time delays at the application level and does not utilize additional traffic over the networks to measure the communication quality. The time-delay compensation technique using the CDOB at the application level is described in this section.

3.1.1 CDOB

As shown in Fig. 5(a), the physical plant with the DOB is

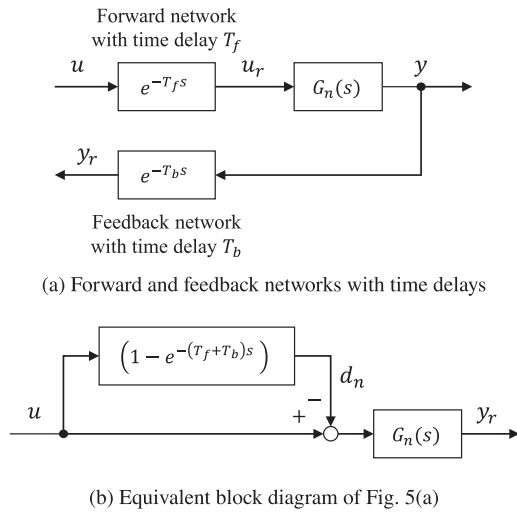


Fig. 5 Time-delayed networks as a disturbance.

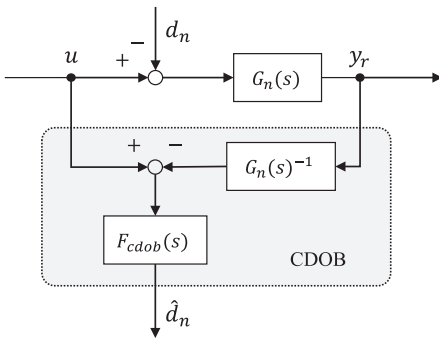


Fig. 6 Network disturbance estimation by the CDOB.

represented as the transfer function $G_n(s)$. In addition, the forward and feedback network delays are defined as T_f and T_b , respectively. The effects of the time delays can be considered as a network disturbance d_n , as shown in Fig. 5(b) and (3)

$$d_n = \left(1 - e^{-(T_f+T_b)s}\right) u. \quad (3)$$

The configuration of the CDOB is shown in Fig. 6. Similar to the DOB, the CDOB estimates the network disturbance without using any time-delay model or measurement [34], [35]. The CDOB comprises the inverse DT model of $G_n(s)$ and the LPF $F_{cdob}(s)$ on the controller side. If the DOB compensates for the local disturbance perfectly, the DT model $G_n(s)$ on the controller side, that is, remote DT, is identical to the characteristics of the physical plant. The estimated network disturbance \hat{d}_n can be calculated as (4)

$$\begin{aligned} \hat{d}_n &= F_{cdob}(s) \left(u - G_n(s)^{-1} y_r\right) \\ &= \frac{g_{cdob}}{s + g_{cdob}} d_n, \end{aligned} \quad (4)$$

where $F_{cdob}(s)$ and g_{cdob} denote the first-order LPF and its cut-off frequency, respectively. The estimated network disturbance \hat{d}_n indicates the effects of the time delays at the

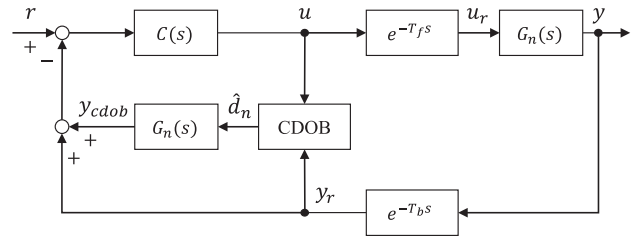


Fig. 7 Time-delay compensation by the CDOB.

application level.

3.1.2 Application-Level Control Using CDOB

In NCSs, the time-delay compensation at the application level means eliminating the time-delay elements, $e^{-T_f s}$ and $e^{-T_b s}$, from the feedback loop. The input-output transfer function of the NCS shown in Fig. 3 is calculated as (5)

$$\frac{y}{r} = \frac{C(s)G_n(s)e^{-T_f s}}{1 + C(s)G_n(s)e^{-(T_f+T_b)s}}, \quad (5)$$

where $C(s)$ denotes the transfer function of the feedback controller. The denominator of the transfer function includes the time delays, T_f and T_b , and the behavior of the system depends on the time-delay elements.

The block diagram of the NCS with the CDOB is shown in Fig. 7. The compensation value y_{cdob} is calculated as (6)

$$y_{cdob} = \frac{g_{cdob}}{s + g_{cdob}} G_n(s) d_n. \quad (6)$$

If the cut-off frequency g_{cdob} is sufficiently large, the input-output transfer function of the NCS is calculated as (7)

$$\frac{y}{r} = \frac{C(s)G_n(s)e^{-T_f s}}{1 + C(s)G_n(s)}, \quad (7)$$

because the network disturbance is cancelled by the output of the CDOB. The denominator of the transfer function does not include the time delays, T_f and T_b , and the feedback controller can be designed without considering the network elements.

3.2 CQEO-Based Estimation and Network Control

The CQEO is an active method to measure the integrated effects of communication quality at the application level by transmitting duplicated data. The path and server selection technique using the CQEOs and redundancy-based network control is described in this section.

3.2.1 CQEO

The original CDOB compensates for the effects of constant time delays in an LTI system, even without being aware of them. However, the CPSs or NCSs are affected by the integrated communication quality, including not only due to constant time delays but also jitters, packet losses, and

where the combination (i, j, k) indicates the path number for i -th forward path and j -th feedback path connected to k -th controller server, and g_{cqe0} denotes the cut-off frequency of the LPF to avoid frequent path switching. As this path selection algorithm is straightforward and not optimized, further studies are required to develop a more sophisticated algorithm. The forward and feedback paths determined by the path selector are utilized to transmit the control input u and response y . The specific algorithms of the path selector for the client-server and peer-to-peer NCSs are described in Sects. 4 and 5, respectively.

4. Path Diversity with CQEO for Client-Server NCSs

This section describes the CQEO-based network control technique using path diversity for the client-server NCSs by taking a remote motor control system as an example.

4.1 Networked Motion Control

A configuration of the remote position servo system is shown in Fig. 10. It is assumed that this system comprises the position controller, communication networks, current-controlled linear motor, and DOB. The position controller is implemented on the server side, and the motor and DOB are implemented on the client side. In Fig. 10, x^{cmd} , x^{res} , and I^{ref} denote the position command, position response, and current reference, respectively. The communication quality disturbances of the forward and feedback networks, $d_{c,f}^I$ and $d_{c,b}^x$, include the effects of time delays, packet losses, and cyberattacks. The received values for I^{ref} and x^{res} over the networks are defined as I^{rec} and x^{rec} , respectively.

If the DOB functions ideally, the transfer function of the motor with the DOB can be represented as (10)

$$G_n(s) = \frac{x^{res}}{I^{ref}} = \frac{K_{tn}}{M_n s^2}, \quad (10)$$

where K_{tn} and M_n denote the nominal thrust constant and nominal mass of the motor, respectively. For example, the position controller can be configured as (11)

$$\begin{aligned} I^{ref} &= \frac{M_n}{K_{tn}} \ddot{x}^{ref} \\ &= \frac{M_n}{K_{tn}} C_p(s) (x^{cmd} - x^{rec}), \end{aligned} \quad (11)$$

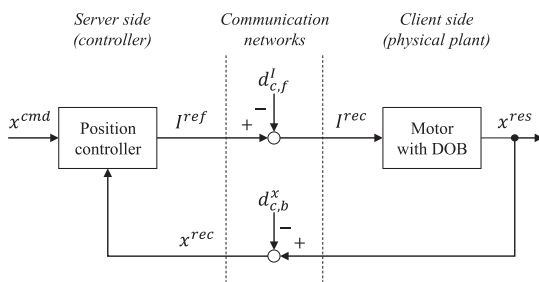


Fig. 10 Remote motor position control system.

where \ddot{x}^{ref} and $C_p(s)$ denote the acceleration reference and the proportional-derivative (PD)-based feedback controller, respectively. The PD-based feedback controller is represented as $C_p(s) = K_p + K_d s$, where K_p and K_d denote the position and velocity feedback gains, respectively.

4.2 System Configuration with CQEO Bank

This study proposes a remote motor position control system using the CQEO-based network control with path diversity. The configuration of the system with two redundant forward and feedback paths is shown in Fig. 11. It is assumed that there is only one controller server in the system, and that there are no server selection algorithms. To improve the reliability of the system, two redundant paths are configured for both forward and feedback networks. The same data are transmitted over the two redundant paths. The communication quality disturbances of forward paths 1 and 2 are defined as $d_{c,f1}^I$ and $d_{c,f2}^x$, respectively. The communication quality disturbances of feedback paths 1 and 2 are defined as $d_{c,b1}^I$ and $d_{c,b2}^x$, respectively.

The path selector determines the combination of forward and feedback paths with the best quality, based on the estimated communication quality disturbance for each path combination (i, j) , where $i \in \{1, 2\}$ and $j \in \{1, 2\}$ denote the path numbers of forward and feedback networks, respectively. The CQEO bank comprises multiple CQEOs for all the path combinations (1, 1), (1, 2), (2, 1), and (2, 2), whose outputs are defined as $\hat{d}_{c,1,1}$, $\hat{d}_{c,1,2}$, $\hat{d}_{c,2,1}$, and $\hat{d}_{c,2,2}$, respectively. The current reference selected for the forward network and position response selected for the feedback network are defined as I^{sel} and x^{sel} , respectively. As shown in Fig. 11, the selected current reference I^{sel} takes a value of either I_1^{rec} or I_2^{rec} , and the selected position response x^{sel} takes a value of either x_1^{rec} or x_2^{rec} .

4.3 CQEO Bank and Path Selector

As described in Sect. 3, each CQEO estimates the communication quality disturbance for each path combination, $\hat{d}_{c,i,j}$. Since the nominal model of the motor is represented as (10), the disturbance $\hat{d}_{c,i,j}$ can be calculated as (12)

$$\hat{d}_{c,i,j} = G_n(s) d_{c,fi}^I + d_{c,bj}^x, \quad (12)$$

where $d_{c,fi}^I$ and $d_{c,bj}^x$ denote the communication quality disturbances for the i -th forward path and j -th feedback path, respectively.

Although various algorithms of the path selector can be adopted, this paper presents an intuitive algorithm based on the absolute values of $\hat{d}_{c,i,j}$. In a similar way to (9), the selected combination of forward and feedback paths is determined as (13)

$$\arg \min_{i \in \{1,2\}, j \in \{1,2\}} \left| \frac{g_{cqe0}}{s + g_{cqe0}} \hat{d}_{c,i,j} \right|, \quad (13)$$

where the combination (i, j) indicates the path number for

\hat{f}_m^{ext} , x_s^{res} , and \hat{f}_s^{ext} over the networks are defined as x_m^{rec} , \hat{f}_m^{rec} , x_s^{rec} , and \hat{f}_s^{rec} , respectively. In practice, the master and slave velocity responses, \dot{x}_m^{res} and \dot{x}_s^{res} , in addition to the position and force responses can be measured. In this case, the velocity response on each side is transmitted over the networks in the same packet as the position and force responses on the corresponding side. The communication quality disturbances and the values received over networks for the velocity response are defined as $d_{c,ms}^{\dot{x}}$ and \dot{x}_m^{rec} on the master-to-slave network, and $d_{c,sm}^{\dot{x}}$ and \dot{x}_s^{rec} on the slave-to-master network, respectively.

In this study, the acceleration-based four-channel bilateral controller [61] is implemented. Current-controlled linear motors with the same nominal mass M_n and nominal thrust constant K_{In} are utilized for actuating the master and slave haptic devices. The DOB is implemented on each haptic device. If the DOB functions ideally, the transfer function of the master and slave haptic devices with the DOB can be represented as (16)

$$G_n(s) = \frac{x_m^{res}}{I_m^{ref}} = \frac{x_s^{res}}{I_s^{ref}} = \frac{K_{In}}{M_n s^2}, \quad (16)$$

where I_m^{ref} and I_s^{ref} denote the current references for the master and slave devices, respectively.

The master and slave controllers can be configured as (17) and (18)

$$I_m^{ref} = \frac{M_n}{K_{In}} \left\{ -C_p(s)(x_m^{res} - x_s^{rec}) - C_f(\hat{f}_m^{ext} + \hat{f}_s^{rec}) \right\}, \quad (17)$$

$$I_s^{ref} = \frac{M_n}{K_{In}} \left\{ -C_p(s)(x_s^{res} - x_m^{rec}) - C_f(\hat{f}_m^{rec} + \hat{f}_s^{ext}) \right\}, \quad (18)$$

where $C_p(s) = K_p + K_d s$ and $C_f = K_f$ are the position controller and force controller, respectively. The gains K_p , K_d , and K_f denote the position, velocity, and force feedback gains, respectively.

If there are no communication quality disturbances and the DOB functions ideally, the transparency in the four-channel bilateral control system can be calculated as (19)

$$\begin{bmatrix} \hat{f}_m^{ext} \\ -x_s^{res} \end{bmatrix} = \begin{bmatrix} -\frac{s^2}{C_f} & 1 \\ -1 & 0 \end{bmatrix} \begin{bmatrix} x_m^{res} \\ -\hat{f}_s^{ext} \end{bmatrix}. \quad (19)$$

Therefore, the perfect transparency, i.e., (14) and (15), is achieved if the force feedback gain K_f is sufficiently large.

5.2 System Configuration with Master and Slave CQEOs

In reality, the bilateral control system is affected by the communication quality disturbances. To achieve reliable haptic communication, communication quality estimation and control techniques are required. This study proposes a haptic communication system using the CQEO-based network control with path diversity. The configuration of the system with

two redundant master-to-slave and slave-to-master paths is shown in Fig. 13. It is assumed that master and slave devices to be utilized are designated in advance, and this system does not have any peer selection algorithms.

To improve the reliability of the system, two redundant paths are configured for both master-to-slave and slave-to-master networks. The same data are transmitted over the two redundant paths. The vector $\mathbf{p}_m^{res} = [x_m^{res}, \dot{x}_m^{res}, \hat{f}_m^{ext}]^T$ is transmitted from the master side to the slave side, and the vector $\mathbf{p}_s^{res} = [x_s^{res}, \dot{x}_s^{res}, \hat{f}_s^{ext}]^T$ is transmitted from the slave side to the master side. The communication quality disturbance vectors of master-to-slave paths 1 and 2 are defined as $\mathbf{d}_{c,ms1} = [d_{c,ms1}^x, d_{c,ms1}^{\dot{x}}, d_{c,ms1}^f]^T$ and $\mathbf{d}_{c,ms2} = [d_{c,ms2}^x, d_{c,ms2}^{\dot{x}}, d_{c,ms2}^f]^T$, respectively. The communication quality disturbance vectors of slave-to-master paths 1 and 2 are defined as $\mathbf{d}_{c,sm1} = [d_{c,sm1}^x, d_{c,sm1}^{\dot{x}}, d_{c,sm1}^f]^T$ and $\mathbf{d}_{c,sm2} = [d_{c,sm2}^x, d_{c,sm2}^{\dot{x}}, d_{c,sm2}^f]^T$, respectively.

In the bilateral control system, the master-to-slave and slave-to-master networks are evaluated separately because the system has two degrees of freedom, that is, two controllers. The master path selector determines the slave-to-master path with the best quality, based on the communication quality disturbance vector $\hat{\mathbf{d}}_{c,m}^x = [\hat{d}_{c,m1}^x, \hat{d}_{c,m2}^x]^T$ estimated by the master CQEO. The slave path selector determines the master-to-slave path with the best quality, based on the communication quality disturbance vector $\hat{\mathbf{d}}_{c,s}^x = [\hat{d}_{c,s1}^x, \hat{d}_{c,s2}^x]^T$ estimated by the slave CQEO. The vectors selected for the master-to-slave and slave-to-master networks are defined as $\mathbf{p}_m^{sel} = [x_m^{sel}, \dot{x}_m^{sel}, \hat{f}_m^{sel}]^T$ and $\mathbf{p}_s^{sel} = [x_s^{sel}, \dot{x}_s^{sel}, \hat{f}_s^{sel}]^T$, respectively. As shown in Fig. 13, the master selected vector \mathbf{p}_m^{sel} takes a vector of either $\mathbf{p}_m^{rec} = [x_{m1}^{rec}, \dot{x}_{m1}^{rec}, \hat{f}_{m1}^{rec}]^T$ or $\mathbf{p}_m^{rec} = [x_{m2}^{rec}, \dot{x}_{m2}^{rec}, \hat{f}_{m2}^{rec}]^T$, and the slave selected vector \mathbf{p}_s^{sel} takes a vector of either $\mathbf{p}_s^{rec} = [x_{s1}^{rec}, \dot{x}_{s1}^{rec}, \hat{f}_{s1}^{rec}]^T$ or $\mathbf{p}_s^{rec} = [x_{s2}^{rec}, \dot{x}_{s2}^{rec}, \hat{f}_{s2}^{rec}]^T$.

5.3 Master and Slave CQEOs and Path Selectors

The master CQEO calculates the estimated disturbances, $\hat{d}_{c,m1}^x$ and $\hat{d}_{c,m2}^x$, as (20) and (21)

$$\hat{d}_{c,m1}^x = x_{s_model}^{res} - x_{s1}^{rec}, \quad (20)$$

$$\hat{d}_{c,m2}^x = x_{s_model}^{res} - x_{s2}^{rec}, \quad (21)$$

where $x_{s_model}^{res}$ denotes the position response of the slave haptic device model on the master side, which includes the position controller $C_p(s)$. The current reference $I_{s_model}^{ref}$ is calculated as (22)

$$I_{s_model}^{ref} = \frac{M_n}{K_{In}} C_p(s)(x_m^{res} - x_{s_model}^{res}). \quad (22)$$

The position response of the slave haptic device model, $x_{s_model}^{res}$, can be calculated as (23)

$$x_{s_model}^{res} = \frac{K_{In}}{M_n s^2} I_{s_model}^{ref}. \quad (23)$$

The slave CQEO calculates the estimated disturbances,

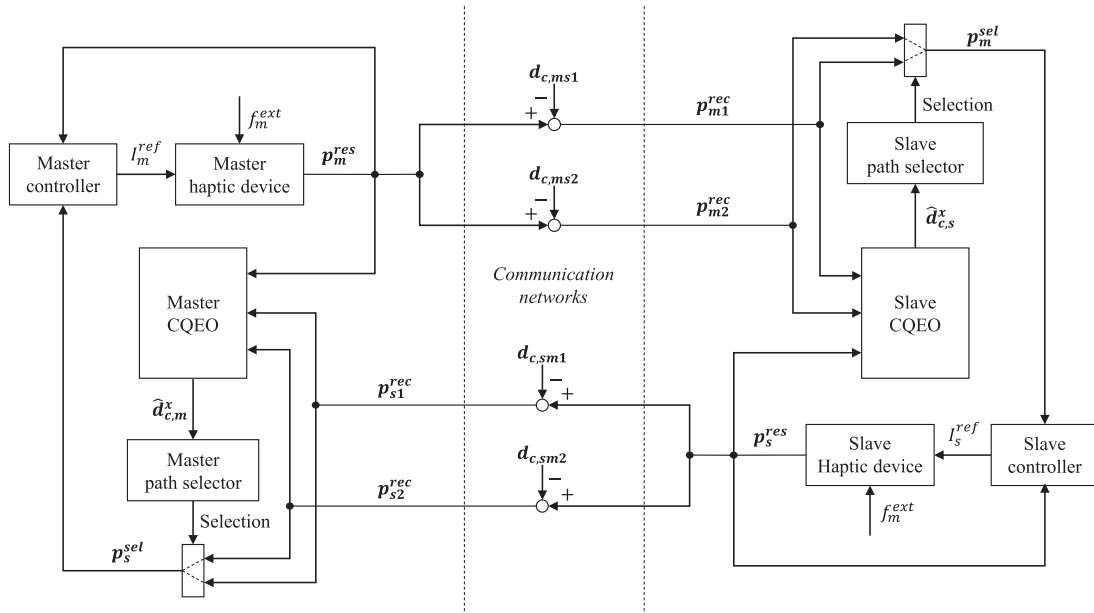


Fig. 13 Haptic communication with path diversity.

$\hat{d}_{c,s1}^x$ and $\hat{d}_{c,s2}^x$, as (24) and (25)

$$\hat{d}_{c,s1}^x = x_{m_model}^{res} - x_{m1}^{rec}, \quad (24)$$

$$\hat{d}_{c,s2}^x = x_{m_model}^{res} - x_{m2}^{rec}, \quad (25)$$

where $x_{m_model}^{res}$ denotes the position response of the master haptic device model on the slave side, which includes the position controller $C_p(s)$. The current reference $I_{m_model}^{ref}$ is calculated as (26)

$$I_{m_model}^{ref} = \frac{M_n}{K_{tn}} C_p(s) (x_s^{res} - x_{m_model}^{res}). \quad (26)$$

The position response of the master haptic device model, $x_{m_model}^{res}$, can be calculated as (27)

$$x_{m_model}^{res} = \frac{K_{tn}}{M_n s^2} I_{m_model}^{ref}. \quad (27)$$

Although various algorithms of the path selectors can be adopted, this study presents an intuitive algorithm based on the absolute values of the communication quality disturbances. In the slave path selector, the selected master-to-slave path is determined as (28)

$$\arg \min_{i \in \{1,2\}} \left| \frac{g_{cqeo}}{s + g_{cqeo}} \hat{d}_{c,si}^x \right|, \quad (28)$$

where i indicates the master-to-slave path number. In the master path selector, the selected slave-to-master path is determined as (29)

$$\arg \min_{j \in \{1,2\}} \left| \frac{g_{cqeo}}{s + g_{cqeo}} \hat{d}_{c,mj}^x \right|, \quad (29)$$

where j indicates the slave-to-master path number.

Table 1 Parameters used in the simulation.

Parameter	Symbol	Value	Unit
Nominal mass	M_n	0.5	kg
Nominal thrust constant	K_{tn}	32.5	N/A
Position feedback gain	K_p	900	
Velocity feedback gain	K_d	60	
Force feedback gain	K_f	1	
Cut-off frequency of the DOB	g_{dob}	500	rad/s
Cut-off frequency of the RFOB	g_{reac}	500	rad/s
Cut-off frequency of the CQEO	g_{cqeo}	1000	rad/s

6. Simulation

This section shows the simulation results with the haptic communication system to confirm the effectiveness of the communication quality estimation and control using the CQEO.

6.1 Setup

In the simulation, the conventional four-channel control system shown in Fig. 12 and the proposed control system with path diversity shown in Fig. 13 were compared. The parameters used in the simulation are shown in Table 1. The reaction force observers (RFOBs) were implemented to estimate the master and slave external forces without force sensors [62]. The RFOBs calculated the external forces, \hat{f}_m^{ext} and \hat{f}_s^{ext} , as (30) and (31)

$$\hat{f}_m^{ext} = \frac{g_{reac}}{s + g_{reac}} f_m^{ext}, \quad (30)$$

$$\hat{f}_s^{ext} = \frac{g_{reac}}{s + g_{reac}} f_s^{ext}, \quad (31)$$

where g_{reac} denotes the cut-off frequency of the first-order LPF. The control period was set to 1 ms. A human operator

Table 2 Network conditions for the system without path diversity.

Cond. No.	Delay	Attack (0 s ≤ t < 10 s)	Attack (10 s ≤ t < 20 s)
1	10 ms	×	✓
2	50 ms	×	✓
3	20–80 ms	×	✓

Table 3 Network conditions for the system with path diversity.

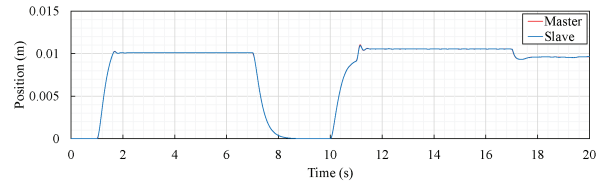
Cond. No.	Path No.	Delay	Attack (0 s ≤ t < 10 s)	Attack (10 s ≤ t < 20 s)
4	1	50 ms	×	×
	2	10 ms	×	✓
5	1	20–80 ms	×	×
	2	50 ms	×	✓

manipulated the master device so that the slave device would contact the remote environment. The mechanical impedance of the human operator Z_h , which indicates the relationship between the master external force f_m^{ext} and master velocity response \dot{x}_m^{res} , was set to $50/s + 10\text{N}\cdot\text{s}/\text{m}$. The position of environment was set to 0.01 m on the slave side. The mechanical impedance of the remote environment Z_e , which indicates the relationship between the slave external force f_s^{ext} and slave velocity response \dot{x}_s^{res} in contact motion, was set to $1000/s + 10\text{N}\cdot\text{s}/\text{m}$.

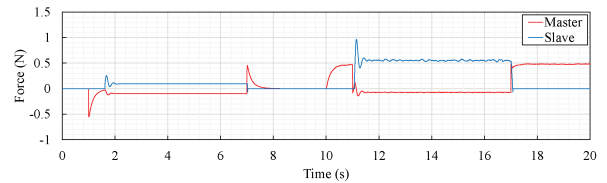
The target of haptic communication is that master and slave position responses take the same value and master and slave force responses follow the law of action and reaction, as shown in (14) and (15). The time delays and data tampering attacks on the master-to-slave and slave-to-master networks were given as the five network conditions shown in Tables 2 and 3, where t denotes the elapsed time. The random jitters followed the uniform distribution in the designated range. The data tampering attacks were injected as an additive disturbance d_a on each position, velocity, and force channel over the master-to-slave and slave-to-master networks. Disturbance d_a was set to a random number ranged from 0 to 0.001 at intervals of 10^{-5} followed uniform distribution. The disturbance is small relative to the range of motion. However, in haptic communication systems, a small displacement can generate a large reaction force when the slave haptic device has contact with a hard object. This is a critical problem if the system is applied to the services involving human lives such as telesurgery. This kind of attacks can be achieved as a man-in-the-middle attack. The simulations assumed that there were no packet losses on the networks. The same network condition was applied to both master-to-slave and slave-to-master networks in the same simulation.

6.2 Results

The simulation results without path diversity for conditions 1, 2, and 3 are shown in Figs. 14–16, respectively. The master and slave position responses under conditions 1, 2, and 3 are shown in Figs. 14(a), 15(a), and 16(a), respectively. The master and slave force responses under conditions 1, 2, and

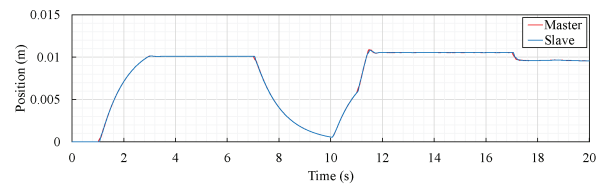


(a) Position responses

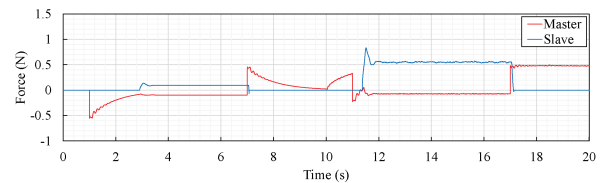


(b) Force responses

Fig. 14 Simulation results without path diversity (Condition 1).

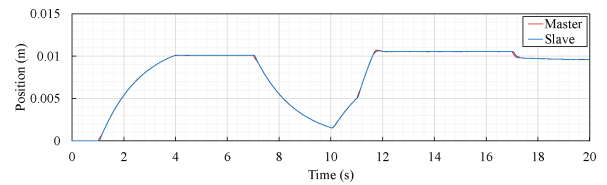


(a) Position responses

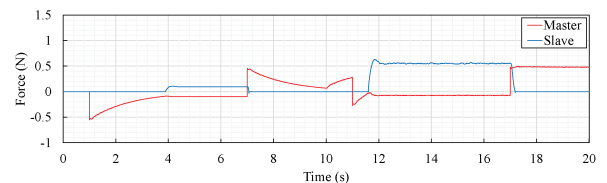


(b) Force responses

Fig. 15 Simulation results without path diversity (Condition 2).



(a) Position responses



(b) Force responses

Fig. 16 Simulation results without path diversity (Condition 3).

3 are shown in Figs. 14(b), 15(b), and 16(b), respectively. In these conditions, only one path was configured for each of master-to-slave and slave-to-master network.

It is clear from Fig. 14 that the position and force responses achieved high-transparency haptic communication, that is, (14) and (15), for the first 10 s while generating the impulsive operational force of approximately -0.5N at 1 s

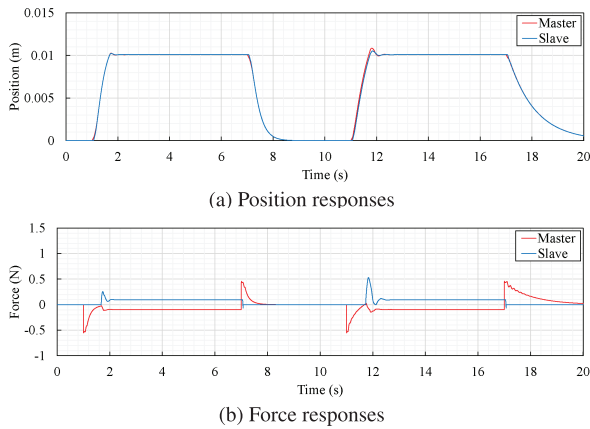


Fig. 17 Simulation results with path diversity (Condition 4).

on the master side. The operational force was generated because the master device did not start moving immediately due to the time delays even if the human operator applied force to the master device. As shown in Fig. 15, the results had slower response times than those of condition 1 because the delay was set to a larger value than that of condition 1. As shown in Fig. 16, the results had slower response times than those of condition 2 because the delay had jitter and the maximum delay was set to 80 ms. In conditions 1–3, after 10 s, the master and slave force responses did not satisfy (15) because of data tampering attacks. In addition, the force responses fluctuated after 10 s, which means that the operator felt the persistent oscillations. Therefore, the systems without path diversity were directly affected by the time delays and data tampering attacks shown in Table 2, because they do not contain any redundant paths.

The simulation results with path diversity for conditions 4 and 5 are shown in Figs. 17 and 18, respectively. The master and slave position responses under conditions 4 and 5 are shown in Figs. 17(a) and 18(a), respectively. The master and slave force responses under conditions 4 and 5 are shown in Figs. 17(b) and 18(b), respectively. In these conditions, two paths were configured for each of master-to-slave and slave-to-master network. In addition, the CQEO and path selector were implemented on each of the master and slave sides.

As confirmed by Figs. 17 and 18, regardless of the presence or absence of the data tampering attacks, the position and force responses achieved high-transparency haptic communication, that is, (14) and (15), while generating the impulsive operational force of approximately -0.5 N at 1 s and 11 s on the master side. As shown in Fig. 17, the results had almost the same response time as those of condition 1 because the path selectors could mainly choose path 2 with a 10-ms delay for the first 10 s. As shown in Fig. 18, the results had almost the same response time as those of condition 2 because the path selectors could mainly choose path 2 with a 50-ms delay for the first 10 s. Therefore, the proposed system with path diversity provided higher transparency than the conventional system without path diversity, because it could

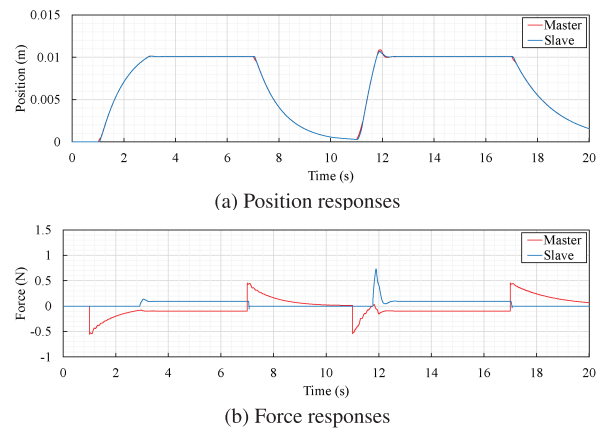


Fig. 18 Simulation results with path diversity (Condition 5).

select an appropriate path and avoid the critical effect of the communication quality disturbance including time delays, jitters, and data tampering attacks based on the estimation of the integrated communication quality at the application level.

7. Conclusion and Future Directions

We described the concept, methodology, and applications of the integrated communication quality estimation and control for the DT-assisted CPSs from both communications and control perspectives. The DT-assisted CPSs were modelled as client-server and peer-to-peer NCSs. Furthermore, we presented the application-level communication quality control using the CDOB and the network-level communication quality control using the CQEO. We also discussed the specific configurations of the CQEO for the client-server and peer-to-peer NCSs by taking the remote motor control and haptic communication systems as examples, respectively. The simulation results showed the effectiveness of the CQEO-based network control technique for the haptic communication system.

In the beyond-5G/6G era, the current MEC-based architecture for delay-sensitive applications will migrate to a co-created multi-access anywhere computing (CMAC) architecture. Distributed computing and network resources will be virtualized and selectively utilized by the applications to achieve the best communication quality. The CMAC architecture provides the communication quality co-created by not only networks and applications but also users, because the CPSs will evolve into cyber-physical-social systems (CPSSs) including human activities [63]. The quality of experiences (QoE) is an index to measure user experiences for network applications [64]. A real-time QoE feedback scheme for network and application control accelerates the research and development of the CMAC-based CPSSs. Therefore, the estimation and control scheme of the integrated communication quality at the human perception level is required to realize future DT-assisted CPSSs.

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